Direct Power Control of a DFIG Fed by a Seven-Level Inverter Using SVM Strategy

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Abstract- In this paper, a direct power control (DPC) is applied to a DFIG (doubly fed induction generator) based wind power conversion systems (WPCSs). The DPC with seven-level space vector modulation (7L-SVM) strategy reduces the harmonic distortion of voltage/current and powers ripples compared with the conventional DPC control scheme; this advantage reduces the ripples of the electromagnetic torque, active and reactive powers. Combined with the DPC technique, the system behavior will present a fast response time, which will guarantee a good tracking of the reference signal. The enhanced performances brought by this strategy are studied in the simulation results.

Keywords: DFIG, DPC, WPCSs, 7L-SVM.

1. Introduction

DPC control scheme using switching table and hysteresis comparators is the traditional strategy used for DFIG-based WPCSs. This strategy is similar to direct torque control (DTC). Major advantages of DPC control scheme are as follows: simple control scheme, fast response time, easy to implement and less dependence on DFIG parameters [1]. The basic idea of the traditional DPC strategy is a direct control of the active and reactive powers of the DFIG without any internal control loop or pulse width modulation technique [2].

In this paper, the DPC strategy with the application of the seven-level SVM strategy has been considered. The original contribution of this paper is the application of the 7L-SVM technique in the DPC control with doubly fed induction generator and simulation investigation of this proposed control scheme.

This paper is divided into seven sections. In Section 1, the introduction is presented. In Section 2, the mathematical model of the DFIG is described. The seven-level SVM technique has been discussed in Section 3. In Section 4, the description of the DPC strategy is presented. Section 5 deals with the description of the DPC strategy with the application of 7L-SVM strategy. simulation studies are presented and discussed in Section 6. The paper is concluded with a short conclusion.

2. The DFIG model

The doubly fed induction generator is described in the Park q-d frame by the following set of equations [3, 4].

Stator flux:

$$\begin{cases} \psi_{ds} = L_s I_{ds} + M I_{dr} \\ \psi_{qs} = L_s I_{qs} + M I_{qr} \end{cases}$$
(1)

Where: Ψ_{qs} and Ψ_{ds} is the stator fluxes. L_s is the inductance of the stator

 I_{ds} and I_{qs} are the stator currents

M is the mutual inductance

Idr and Iqr are the rotor currents Rotor flux:

$$\begin{cases} \psi_{dr} = L_r I_{dr} + M I_{ds} \\ \psi_{qr} = L_r I_{qr} + M I_{qs} \end{cases}$$

$$\tag{2}$$

Where: Ψ_{dr} and Ψ_{qr} is the rotor fluxes

L_r is the inductance of the rotor Stator voltage:

$$\begin{cases} V_{ds} = R_s I_{ds} + \frac{d}{dt} \psi_{ds} - \omega_s \psi_{qs} \\ V_{qs} = R_s I_{qs} + \frac{d}{dt} \psi_{qs} + \omega_s \psi_{ds} \end{cases}$$
(3)

Where: V_{ds} and V_{qs} are the stator Voltages R_s is the stator resistance

 ω_s : is the electrical pulsation of the stator. Rotor voltage:

$$\begin{cases} V_{dr} = R_r I_{dr} + \frac{d}{dt} \psi_{dr} - \omega_r \psi_{qr} \\ V_{qr} = R_r I_{qr} + \frac{d}{dt} \psi_{qr} + \omega_r \psi_{dr} \end{cases}$$
(4)

Where : V_{dr} and V_{qr} are the rotor voltages

R_r is the rotor resistance

The stator side reactive and active powers are defined as:

$$\begin{cases} P_{s} = \frac{3}{2} (V_{ds} I_{ds} + V_{qs} I_{qs}) \\ Q_{s} = \frac{3}{2} (V_{qs} I_{ds} - V_{ds} I_{qs}) \end{cases}$$
(5)

Where: P_s is the active power

Q_s is the reactive power.

The electrical model of the DFIG is completed by the following mechanical equation:

$$T_e - T_r = J \cdot \frac{d\Omega}{dt} + f \cdot \Omega \tag{6}$$

Where: Ω is the mechanical rotor speed.

T_e is the electromagnetic torque

J is the inertia

f is the viscous friction coefficient

 T_r is the load torque.

The electromagnetic torque can be written as follows:

$$T_{e} = \frac{3}{2} p \frac{M}{L_{r}} (I_{dr} \cdot \psi_{qs} - I_{qr} \cdot \psi_{ds})$$
(7)

Where: *p* is the number of pole pairs.

3.Seven-level SVM strategy

Traditionally the SVM technique is widely used in control of machine drives. In this strategy is need to calculate the sector and angle. In addition, this strategy is difficult to implement. In [5], the author propose an new SVM algorithm based on calculating the maximum (Max) and minimum (Min) of three-phase voltages. This proposed SVM method is simple and easy to implement compared to classical SVM strategy. In [6], the author propose a five-level SVM strategy to control the inverter of DFIG. The SVM strategy of the four-level inverter is proposed in [7]. This paper presents a SVM modulation scheme of seven-level NPC inverter. Fig. 1 shows the SVM technique of seven-level NPC inverter. It is observed that the proposed strategy is simple control and easy to implement. This proposed SVM strategy can be minimized the harmonic distortion of the voltage/current and electromagnetic torque ripple of the DFIG-based WPCSs. On the other hand, the seven-level SVM strategy based on the following steps:



Fig. 1. Block diagram of seven-level SVM strategy.

- Calculates the minimum voltages (min (Va, Vb, Vc))
- Calculates the maximum voltages (max (Va, Vb, Vc))
- Add the maximum and minimum voltages (max (Va, Vb, Vc) + min (Va, Vb, Vc)).

• The last step is to compare step-3 waveforms with Vp1 ($V_{Triangle}$) to Vp6, and generates the pulses for that switch presents in the three-phase voltage source converter circuit.

4. Classical DPC control

The DPC goal is to control the reactive and active powers of the DFIG-based wind turbine systems (WTSs). In this strategy the active and reactive powers that are directly controlled [8]. The traditional DPC, which is designed to control stator reactive and active powers of the DFIG, is shown in Fig. 2 [9].



Fig. 2. Conventional DPC strategy.

Rotor flux is estimated using (8) and (10) [10].

$$\Psi_{r\beta} = \sigma L_r I_{r\beta} \tag{8}$$

Where : $\Psi_{r\beta}$: is the rotor flux linkage of β -axis. I_{r β} : is the rotor current linkage of β -axis. And :

$$\sigma = 1 - \frac{M^2}{L_r L_s} \tag{9}$$

$$\Psi_{r\alpha} = \sigma L_r I_{r\alpha} + \frac{M}{L_s} \Psi_s \tag{10}$$

Where : $\Psi_{r\alpha}$: is the rotor flux linkage of α -axis. Ψ_s is the stator flux.

 $I_{r\alpha}$: is the rotor current linkage of α -axis.

The stator flux amplitude is given by:

$$\Psi_s = \sqrt{\Psi_{s\alpha}^2 + \Psi_{s\beta}^2} \tag{11}$$

Where: $\Psi_{s\alpha}$: is the stator flux linkage of α -axis.

 $\Psi_{s\beta}$: is the stator flux linkage of β -axis.

And :

$$\left|\overline{\Psi_s}\right| = \frac{\left|V_s\right|}{w_s} \tag{12}$$

Where : V_s is the stator voltage.

Reactive and active powers is estimated using (13) and (14) [11, 12].

$$Q_{s} = -\frac{3}{2} \left(\frac{V_{s}}{\sigma.L_{s}} \cdot \psi_{r\beta} - \frac{V_{s.Lm}}{\sigma.L_{s.Lr}} \cdot \psi_{r\alpha} \right)$$
(13)

$$P_{s} = -\frac{3}{2} \frac{L_{m}}{\sigma.L_{s}.L_{r}} \cdot (V_{s}.\psi_{r\beta})$$
(14)

Where: Lm is the mutual inductance

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In conventional DPC control scheme a three-level hysteresis comparator (Fig. 4) is used for the stator active power (Hp) and a two-level hysteresis comparator (Fig. 5) for the reactive power (Hq). On the other hand, based on the values of constants Hp and Hq and the position of the rotor flux, the inverter switching algorithm is as shown in Table 1.



Fig. 3. Reactive power hysteresis comparator.



Fig. 4. Active power hysteresis comparator.

Table 1. Switching table for traditional DPC strategy

Ν		1	2	3	4	5	6
Hq	Нр						
	1	5	6	1	2	3	4
1	0	7	0	7	0	7	0
	-1	3	4	5	6	1	2
	1	6	1	2	3	4	5
0	0	0	7	0	7	0	7
	-1	2	3	4	5	6	1

5. DPC with 7L-SVM strategy

In order to improve the traditional DPC control performances a complimentary use of seven-level SVM strategy is proposed. This proposed control scheme reduced harmonic distortion of stator voltage/current, electromagnetic torque ripples, reactive and active ripples compared to classical DPC strategy. This proposed control is simple and easy to implement. The proposed seven-level DPC-SVM control scheme designed to control the stator active and reactive powers of the DFIG system is illustrated in Fig. 5.

The PI controller gains (Ki and Kp) were found after performing simulations in Matlab/Simulink software. Table 2 shows the constants values.

Tahle	2	ΡI	controller	gaines
I abie	4.	11	controller	games

Ki	0.6
Кр	0.00005

6.Simulation results

To study the performance of the proposed control strategies, seven-level DPC-SVM and classical DPC control scheme, the simulation of the system was conducted using Matlab/Simulink software.



Fig. 5. Block diagram of seven-level DPC-SVM control.

MW, 380/690V,

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two poles, 50Hz, with the following parameters : $Rs = 0.012\Omega$, $Rr = 0.021\Omega$, Ls = 0.0137 H, Lr = 0.0136 H, M = 0.0135 H.

The system has the following mechanical parameters : $J = 1000 \text{ Kg.m}^2$, $f_r = 0.0024 \text{ Nm/s}$.

A. Reference tracking test (RTT)

From the simulation results presented in Fig. 6 and Fig. 7 it is apparent that the harmonic distortion (THD) of stator current for the seven-level DPC-SVM control is considerably minimized compared to classical DPC control scheme. Table 3 shows the comparative analysis of THD value for stator current.

Table 3. Comparative analysis of THD value

DPC	Seven-level DPC-SVM
1%	0.39%

From the system responses given in Figs. 8-9 for DPC control and seven-level DPC-SVM strategy the reactive and active powers tracks the reference powers without overshoot, with zero steady state error. Fig. 10 shows the electromagnetic torque of DPC and proposed strategy.

From Figs. 11-13 can be seen that the proposed strategy minimized the torque ripples, stator active and reactive powers ripples compared to classical DPC strategy.



Fig. 6. Spectrum harmonic of rotor current (classical DPC).



Fig. 7. Spectrum harmonic of rotor current (Seven-level DPC-SVM)



Fig. 8. Active power (RTT).

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Fig. 13. Zoom in the torque (RTT).

B.Robustness Test(RT)

In order to test the robustness of the proposed strategies, the DFIG parameters have been intentionally changed such as the values of the resistances R_s and R_r are multiplied by 2 and the values of the inductances L_s and L_r are divided by 2. Simulation results are presented in Figs. 14-18. Figs.14-15 shows the THD of stator current for the proposed strategies. It is found that seven-level DPC-SVM control reduced the THD compared to classical DPC (See Table 4).

Table 4. Comparative analysis of THD value

DPC	Seven-level DPC-SVM
1.26%	0.70%

Fig. 16 shows the active power responses of both the classical DPC control and seven-level DPC-SVM control scheme. It is found that the seven-level DPC-SVM exhibits smooth response and lesser ripples in active power as compared to the classical DPC strategy. Fig. 17 shows the zoom of the reactive power for classical DPC and proposed control scheme respectively. It is shown that the proposed strategy gives minimum reactive power ripples compared to classical DPC.

Electromagnetic torque response comparing curves are shown in Fig. 18. It can be seen that the torque ripple is significantly minimized when the seven-level DPC-SVM is in use. Thus, it can be concluded that the proposed sevenlevel DPC-SVM control scheme is more robust than the classical DPC one.



Fig. 14. Spectrum harmonic of rotor current (classical DPC)

Frequency (Hz)



Fig. 15. Spectrum harmonic of rotor current (Seven-level DPC-SVM)

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b)Seven-level DPC-SVM Fig. 21. Zoom in the active stator power (RT).

7.Conclusion

A seven-level direct power control with space vector modulation for a wind power conversion systems equipped with DFIG controlled through power converter has been proposed in this work. This proposed control scheme is used to reduces the harmonic distortion of current/voltage and powers ripples of DFIG-based WTSs. The seven-level DPC-SVM control scheme is a good candidate for controlling the WPCSs based on a DFIG interconnected to the grid.

Appendix

Table 5. Wind turbine parameters

Blade radius	R	35.25m
Number of blades	-	3
Gearbox ratio	G	90
Moment of inertia	J	1000 Kg.m ²
Viscous friction coefficient	fr	0.0024 N.m/s
Cut-in wind speed	-	4 m/s
Cut-out wind speed	-	m/s
Nominal wind speed	v	m/s

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